

**48Vin  
Input**

**1.2Vout  
Output**

**60Amp  
Current**

**2000Vdc  
Isolation**

**Quarter-brick  
DC/DC Converter**

The PQ60012QPA60 PowerQor® Peta quarter-brick converter is a next-generation, board-mountable, isolated, fixed switching frequency DC/DC converter that uses synchronous rectification to achieve extremely high conversion efficiency. The power dissipated by the converter is so low that a heatsink is not required, which saves cost, weight, height, and application effort. All of the power and control components are mounted to the multi-layer PCB substrate with high-yield surface mount technology. Since the PowerQor converter has no explicit thermal interfaces, it is extremely reliable. The Peta series offers industry leading output currents for any standard quarter-brick module. RoHS Compliant (see page 12).

## PowerQor® Peta



*PQ60012QPA60 Module*

### Operational Features

- Ultra-high efficiency, 87% half load, 84% full load
- Delivers up to 60 amps of output current with minimal derating - no heatsink required
- Wide input voltage range: 36V – 75V, with 100V 100ms input voltage transient capability
- Fixed frequency switching provides predictable EMI performance
- No minimum load requirement means no preload resistors required

### Mechanical Features

- Industry standard quarter-brick pin-out configuration
- Industry standard size: 1.45" x 2.3" (36.8x58.4mm)
- Total height less than 0.45" (11.4mm), permits better airflow and smaller card pitch
- Total weight: 1.5 oz. (42 grams)
- Flanged pins designed to permit surface mount soldering (avoid wave solder) using FPIP technique

### Control Features

- On/Off control referenced to input side (positive and negative logic options are available)
- Remote sense for the output voltage compensates for output distribution drops
- Output voltage trim permits custom voltages and voltage margining

### Protection Features

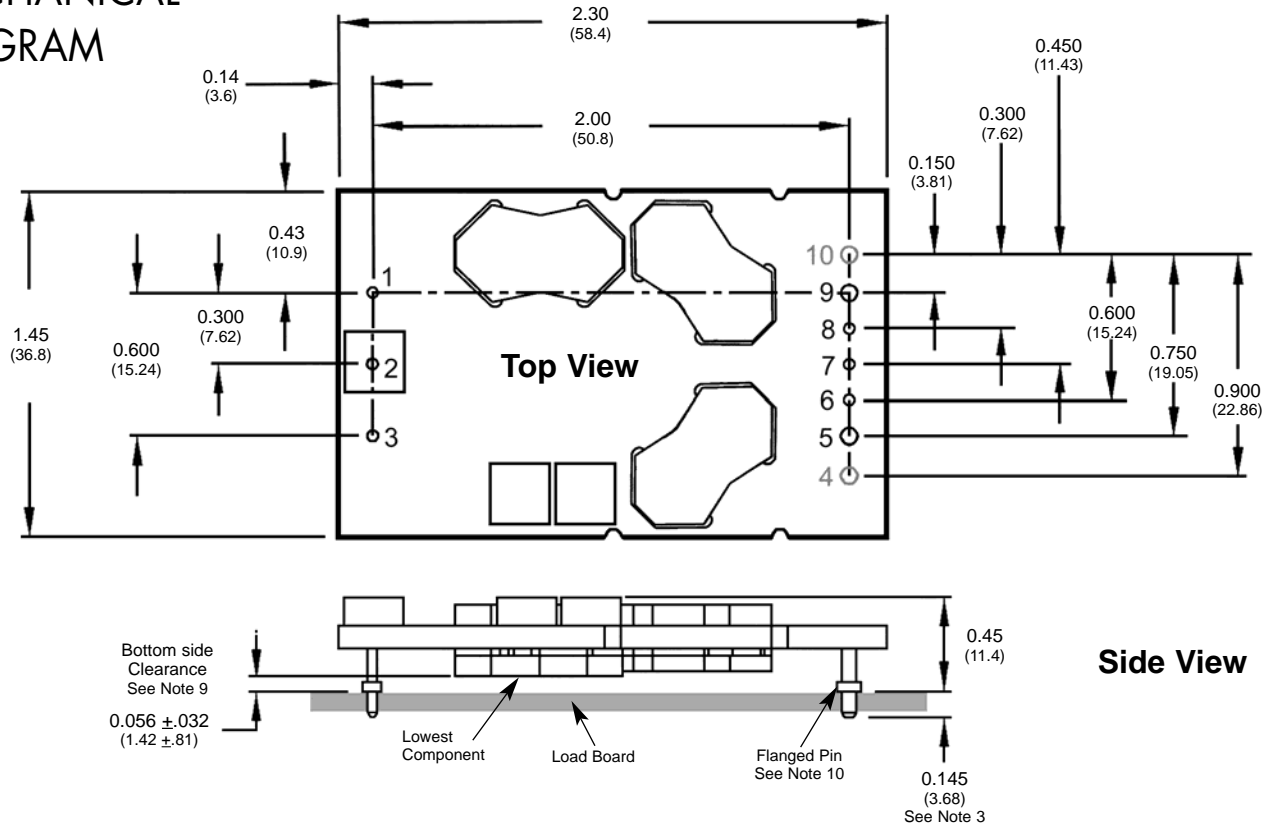
- Input under-voltage lockout disables converter at low input voltage conditions
- Output current limit and short circuit protection protects converter and load from damage and consequent hazardous conditions
- Active back bias limit prevents damage to converter from external load induced pre-bias
- Output over-voltage protection protects load from damaging voltages
- Thermal shutdown protects converter from abnormal environmental conditions

### Safety Features

- 2000V, 30 MΩ input-to-output isolation provides input/output ground separation
- UL/cUL 60950-1 recognized (US & Canada), basic insulation rating
- TUV certified to EN60950-1
- Meets 72/23/EEC and 93/68/EEC directives which facilitates CE Marking in user's end product
- Board and plastic components meet UL94V-0 flammability requirements

**Input: 36-75 V**  
**Output: 1.2 V**  
**Current: 60 A**  
**Package: Quarter-brick**

### MECHANICAL DIAGRAM



### NOTES

- 1) Pins 1-3, 6-8 are 0.040" (1.02mm) diameter with 0.080" (2.03 mm) diameter standoff shoulders.
- 2) Pins 4-5, 9-10 are 0.062" (1.57 mm) diameter with 0.100" (2.54 mm) diameter standoff shoulders.
- 3) Other pin extension lengths available. Recommended pin length is 0.03" (0.76mm) greater than the PCB thickness.
- 4) All Pins: Material - Copper Alloy  
 Finish (RoHS 5/6) - Tin/Lead over Nickel plate  
 Finish (RoHS 6/6) - Matte Tin over Nickel plate
- 5) Undimensioned components are shown for visual reference only.
- 6) All dimensions in inches (mm)  
 Tolerances: x.xx +/-0.02 in. (x.x +/-0.5mm)  
 x.xxx +/-0.010 in. (x.xx +/-0.25mm)
- 7) Weight: 1.5 oz. (42 g) typical
- 8) Workmanship: Meets or exceeds IPC-A-610C Class II
- 9) UL/TUV standards require a clearance greater than 0.04" (1.02mm) between input and output for Basic insulation. This issue should be considered if any copper traces are on the top side of the user's board. Note that the ferrite cores are considered part of the input/primary circuit.
- 10) The flanged pins are designed to permit surface mount soldering (allowing to avoid the wave soldering process) through the use of the flanged pin-in-paste technique.

### PIN DESIGNATIONS

Pin No.	Name	Function
1	Vin(+)	Positive input voltage
2	ON/OFF	TTL input to turn converter on and off, referenced to Vin(-), with internal pull up.
3	Vin(-)	Negative input voltage
4	<i>Vout(+)</i>	<i>Positive output voltage</i>
5	Vout(-)	Negative output voltage
6	SENSE(-)	Negative remote sense <sup>1</sup>
7	TRIM	Output voltage trim <sup>2</sup>
8	SENSE(+)	Positive remote sense <sup>3</sup>
9	Vout(+)	Positive output voltage
10	<i>Vout(-)</i>	<i>Negative output voltage</i>

*Pins in Italics Shaded text are Optional*

#### Notes:

1. SENSE(-) should be connected to Vout(-) either remotely or at the converter.
2. Leave TRIM pin open for nominal output voltage.
3. SENSE(+) should be connected to Vout(+) either remotely or at the converter.

## PQ60012QPA60 ELECTRICAL CHARACTERISTICS

T<sub>A</sub>=25°C, airflow rate=300 LFM, V<sub>in</sub>=48Vdc unless otherwise noted; full operating temperature range is -40°C to +100°C ambient temperature with appropriate power derating. Specifications subject to change without notice.

Parameter	Min.	Typ.	Max.	Units	Notes & Conditions
<b>ABSOLUTE MAXIMUM RATINGS</b>					
Input Voltage					
Non-Operating			100	V	Continuous
Operating			80	V	Continuous
Operating Transient Protection			100	V	100ms transient, square wave
Isolation Voltage (input to output)			2000	V	Basic insulation, Pollution Degree 2
Operating Temperature	-40		100	°C	
Storage Temperature	-55		125	°C	
Voltage at ON/OFF input pin	-2		18	V	
<b>INPUT CHARACTERISTICS</b>					
Operating Input Voltage Range	36	48	75	V	
Input Under-Voltage Lockout					
Turn-On Voltage Threshold	32	33	34	V	
Turn-Off Voltage Threshold	28.5	29.5	30.5	V	
Lockout Voltage Hysteresis	2.5	3.5	4.5	V	
Maximum Input Current			3.0	A	100% Load, 36 Vin
No-Load Input Current		50	100	mA	
Disabled Input Current		2	4	mA	
Inrush Current Transient Rating			0.03	A <sup>2</sup> s	
Response to Input Transient		200		mV	1000V/ms input transient
Input Reflected Ripple Current		50		mA	P-P thru 10µH inductor; Figures 13 & 15
Input Terminal Ripple Current		85		mA	RMS; Figures 13 & 14
Recommended Input Fuse			20	A	Fast blow external fuse recommended
Input Filter Component Values (CVL)		4.4\2.0		µF\µH	Internal values; see Figure F
Recommended External Input Capacitance		47		µF	Typical ESR 0.1-0.2Ω; see Figure 13
<b>OUTPUT CHARACTERISTICS</b>					
Output Voltage Set Point	1.180	1.200	1.220	V	
Output Voltage Regulation					
Over Line		±0.2 \ 2	±0.3 \ 4	%\mV	
Over Load		±0.2 \ 2	±0.3 \ 4	%\mV	
Over Temperature		±5	±16	mV	
Total Output Voltage Range	1.160		1.240	V	Over sample, line, load, temperature & life
Output Voltage Ripple and Noise <sup>1</sup>					500MHz bandwidth; see Figures 13 & 16
Peak-to-Peak		72	120	mV	Full Load; see Figures 13 & 16
RMS		15	25	mV	Full Load; see Figures 13 & 16
Operating Output Current Range	0		60	A	Subject to thermal derating; Figures 5 - 8
Output DC Current Limit Inception	63	70	77	A	Output Voltage 10% Low
Output DC Current Limit Shutdown Voltage		0.8		V	
Back-Drive Current Limit while Enabled	0.15	2	5.0	A	Negative current drawn from output
Back-Drive Current Limit while Disabled	0	10	50	mA	Negative current drawn from output
Maximum Output Capacitance			30,000	µF	1.2Vout at 60A Resistive Load
<b>DYNAMIC CHARACTERISTICS</b>					
Input Voltage Ripple Rejection		90		dB	120 Hz; Figure 20
Output Voltage during Load Current Transient					
For a Step Change in Output Current (0.1A/µs)		80		mV	50% to 75% to 50% lout max; Figure 11
For a Step Change in Output Current (5A/µs)		80		mV	50% to 75% to 50% lout max; Figure 12
Settling Time		400		µs	To within 1% Vout nom
Turn-On Transient					
Turn-On Time		15	20	ms	Full load, Vout=90% nom.; Figures 9 & 10
Start-Up Inhibit Time	180	200	240	ms	-40°C to +125°C; Figure G
Output Voltage Overshoot		0		%	10,000 µF load capacitance, lout = 0A
<b>EFFICIENCY</b>					
100% Load	82	84		%	Figures 1 - 4
50% Load	85	87		%	Figures 1 - 4
<b>TEMPERATURE LIMITS FOR POWER DERATING CURVES</b>					
Semiconductor Junction Temperature			125	°C	Package rated to 150°C
Board Temperature			125	°C	UL rated max operating temp 130°C
Transformer Temperature			125	°C	See Figures 5 - 8 for derating curves
<b>ISOLATION CHARACTERISTICS</b>					
Isolation Voltage (dielectric strength)		2000		V	
Isolation Resistance		30		MΩ	
Isolation Capacitance <sup>2</sup>			470	pF	

Note 1: For applications requiring reduced output voltage ripple and noise, consult SynQor applications support (e-mail: [support@synqor.com](mailto:support@synqor.com))

Note 2: Higher values of isolation capacitance can be added external to the module.

## ELECTRICAL CHARACTERISTICS (Continued)

Parameter	Min.	Typ.	Max.	Units	Notes & Conditions
<b>FEATURE CHARACTERISTICS</b>					
Regulation Stage Switching Frequency	250	275	300	kHz	
Isolation Stage Switching Frequency		137.5		kHz	
ON/OFF Control (Option P)					
Off-State Voltage	-2		0.8	V	
On-State Voltage	2.4		18	V	
ON/OFF Control (Option N)					
Off-State Voltage	2.4		18	V	
On-State Voltage	-2		0.8	V	
ON/OFF Control (Either Option)					Figures <a href="#">A</a> & <a href="#">B</a>
Pull-Up Voltage		$V_{in}/6.5$	9.2	V	
Pull-Up Resistance		40		k $\Omega$	
Output Voltage Trim Range	-10		+10	%	Measured across Pins 10 & 4; Figure <a href="#">C</a>
Output Voltage Remote Sense Range			+10	%	Measured across Pins 10 & 4
Output Over-Voltage Protection	1.53	1.61	1.68	V	Over full temp range; % of nominal $V_{out}$
Over-Temperature Shutdown		125		$^{\circ}$ C	Average PCB Temperature
Over-Temperature Shutdown Restart Hysteresis		10		$^{\circ}$ C	
Load Current Scale Factor		2400			See App Note: Output Load Current Calc.
<b>RELIABILITY CHARACTERISTICS</b>					
Calculated MTBF (Telcordia)		2.37		10 <sup>6</sup> Hrs.	TR-NWT-000332; 80% load, 300LFM, 40 $^{\circ}$ C $T_a$
Calculated MTBF (MIL-217)		1.75		10 <sup>6</sup> Hrs.	MIL-HDBK-217F; 80% load, 300LFM, 40 $^{\circ}$ C $T_a$
Field Demonstrated MTBF				10 <sup>6</sup> Hrs.	See <a href="#">website</a> for details

## STANDARDS COMPLIANCE

Parameter	Notes
<b>STANDARDS COMPLIANCE</b>	
UL/cUL 60950-1	File # E194341, Basic insulation & pollution degree 2
EN60950-1	Certified by TUV
72/23/EEC	
93/68/EEC	
Needle Flame Test (IEC 695-2-2)	Test on entire assembly; board & plastic components UL94V-0 compliant
IEC 61000-4-2	ESD test, 8kV - NP, 15kV air - NP (Normal Performance)
GR-1089-CORE	Section 7 - electrical safety, Section 9 - bonding/grounding
Telcordia (Bellcore) GR-513	

- An external input fuse must always be used to meet these safety requirements. Contact SynQor for official safety certificates on new releases or download from the SynQor website.

## QUALIFICATION TESTING

Parameter	# Units	Test Conditions
<b>QUALIFICATION TESTING</b>		
Life Test	32	95% rated $V_{in}$ and load, units at derating point, 1000 hours
Vibration	5	10-55Hz sweep, 0.060" total excursion, 1 min./sweep, 120 sweeps for 3 axis
Mechanical Shock	5	100g minimum, 2 drops in x and y axis, 1 drop in z axis
Temperature Cycling	10	-40 $^{\circ}$ C to 100 $^{\circ}$ C, unit temp. ramp 15 $^{\circ}$ C/min., 500 cycles
Power/Thermal Cycling	5	Toperating = min to max, $V_{in}$ = min to max, full load, 100 cycles
Design Marginality	5	$T_{min}$ -10 $^{\circ}$ C to $T_{max}$ +10 $^{\circ}$ C, 5 $^{\circ}$ C steps, $V_{in}$ = min to max, 0-105% load
Humidity	5	85 $^{\circ}$ C, 85% RH, 1000 hours, 2 minutes on and 6 hours off
Solderability	15 pins	MIL-STD-883, method 2003

- Extensive characterization testing of all SynQor products and manufacturing processes is performed to ensure that we supply robust, reliable product. Contact the factory for official product family qualification documents.

## OPTIONS

SynQor provides various options for Logic Sense, Pin Length and Feature Set for this family of DC/DC converters. Please consult the [last page](#) of this specification sheet for information on available options.

## PATENTS

SynQor is protected under various patents. Please consult the [last page](#) for further information.

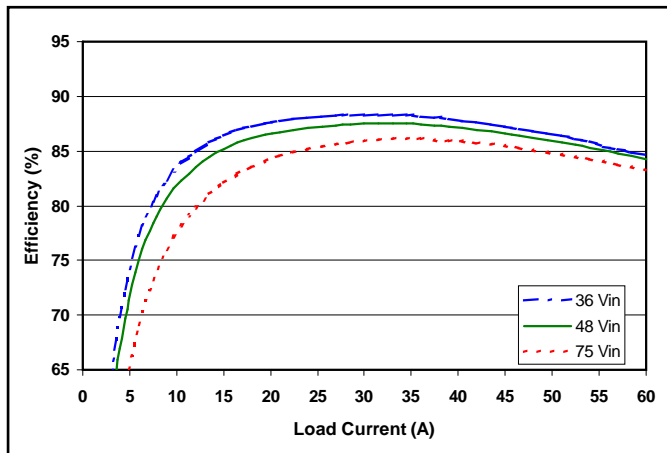


Figure 1: Efficiency at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

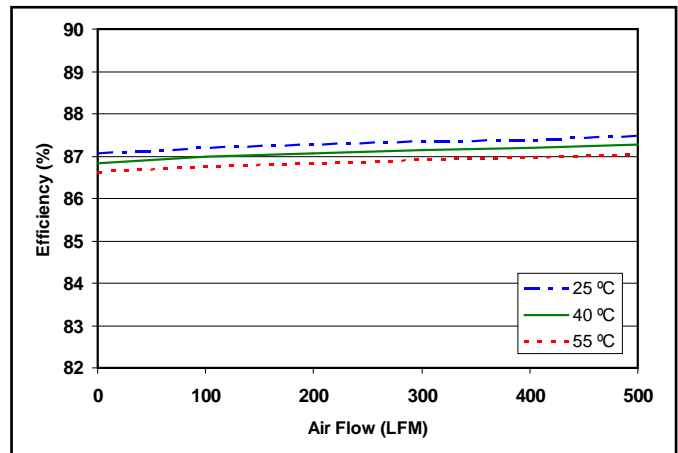


Figure 2: Efficiency at nominal output voltage and 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C (nominal input voltage).

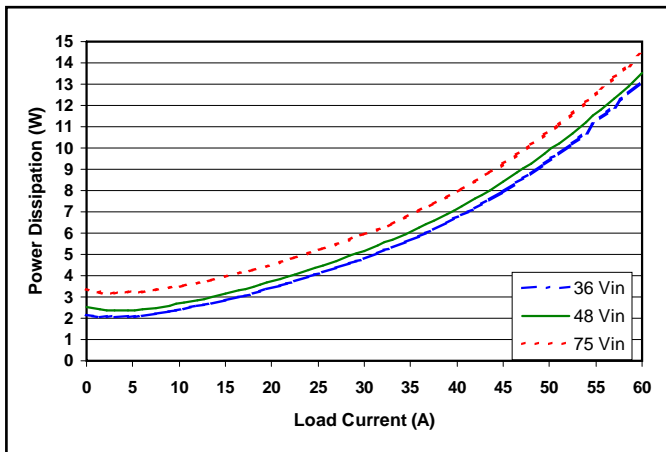


Figure 3: Power dissipation at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

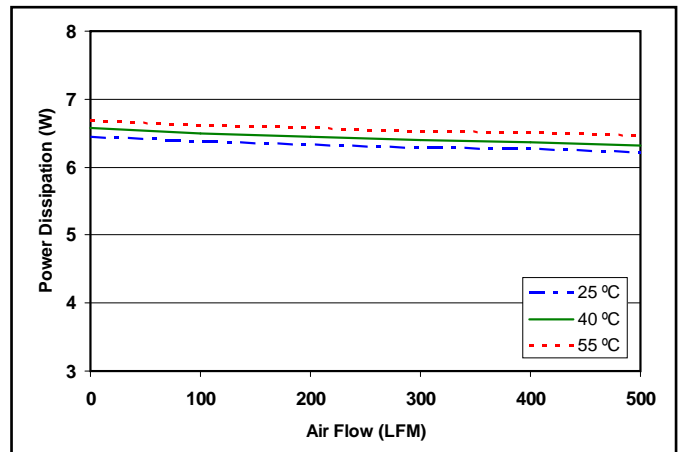


Figure 4: Power dissipation at nominal output voltage and 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C (nominal input voltage).

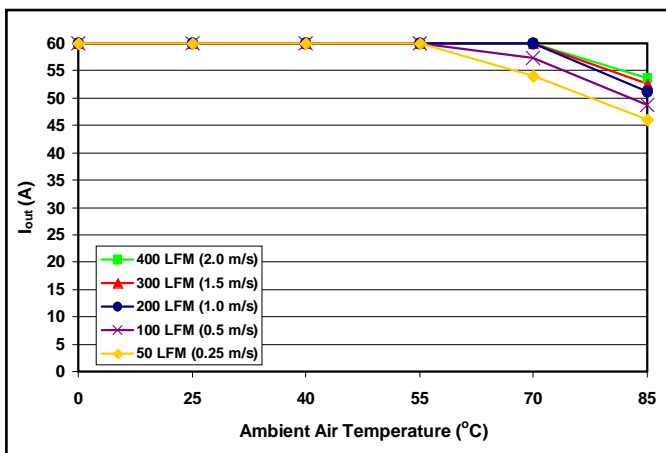


Figure 5: Maximum output power derating curves vs. ambient air temperature for airflow rates of 50 LFM through 400 LFM with air flowing across the converter from pin 3 to pin 1 (nominal input voltage).

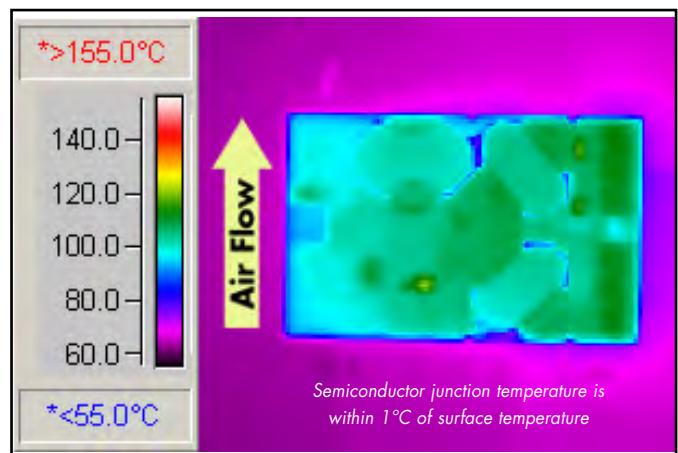
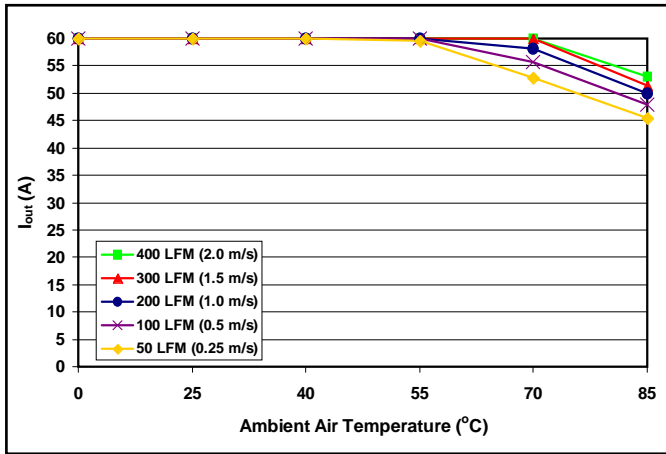
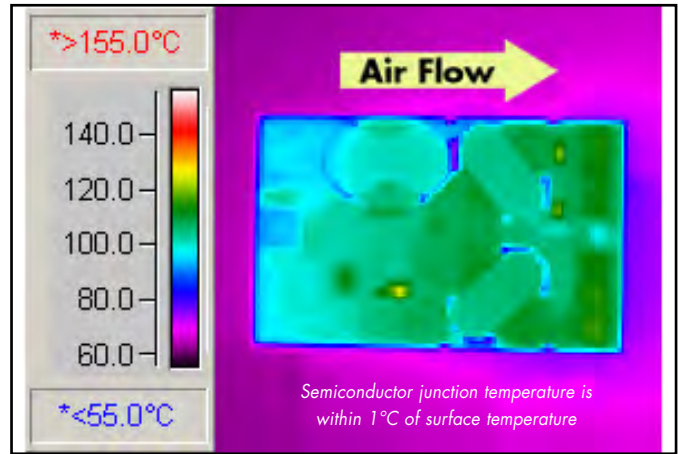


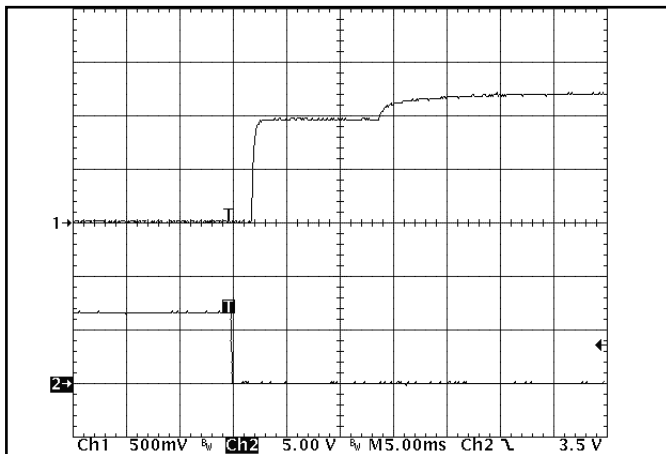
Figure 6: Thermal plot of converter at 60 amp load current with 55°C air flowing at the rate of 200 LFM. Air is flowing across the converter from pin 3 to pin 1 (nominal input voltage).



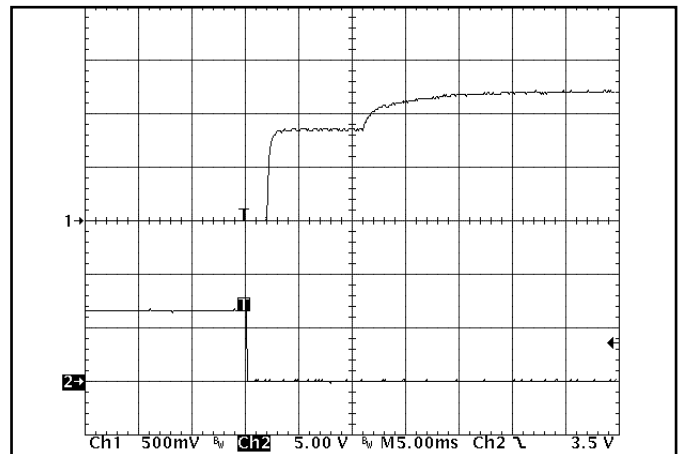
**Figure 7:** Maximum output power derating curves vs. ambient air temperature for airflow rates of 50 LFM through 400 LFM with air flowing from input to output (nominal input voltage).



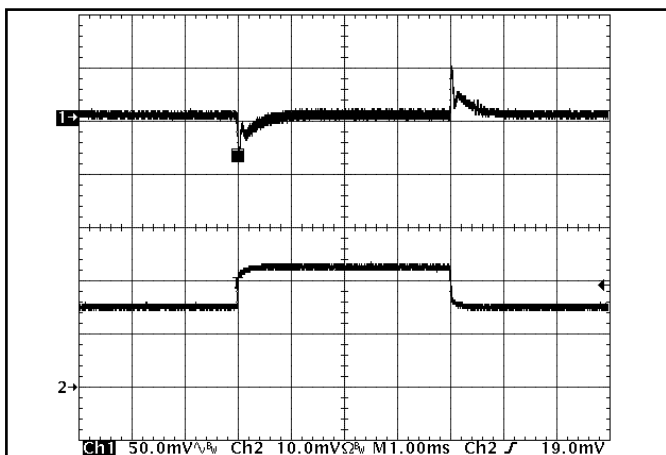
**Figure 8:** Thermal plot of converter at 60 amp load current with 55°C air flowing at the rate of 200 LFM. Air is flowing across the converter from input to output (nominal input voltage).



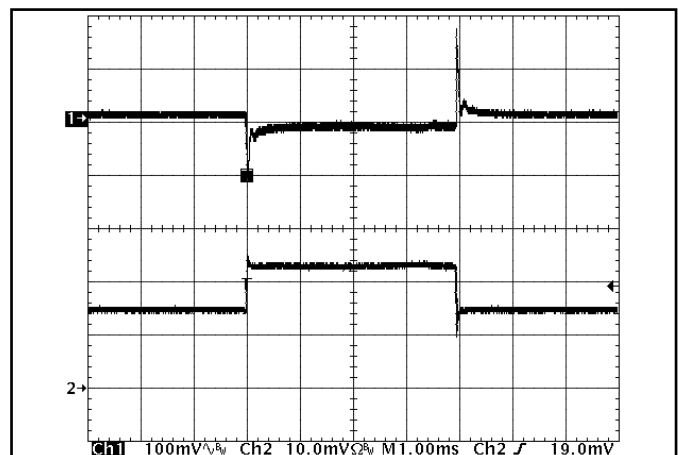
**Figure 9:** Turn-on transient at full rated load (resistive load) (5 ms/div). Input voltage pre-applied. Ch 1:  $V_{out}$  (500mV/div), Ch 2: ON/OFF input (5V/div).



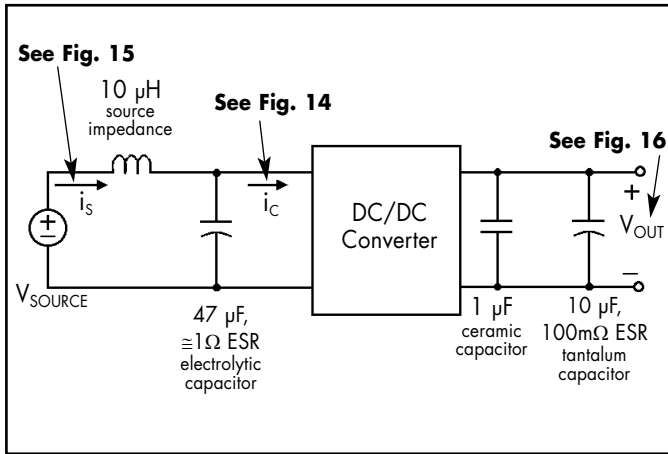
**Figure 10:** Turn-on transient at zero load current (5 ms/div). Ch 1:  $V_{out}$  (500mV/div), Ch 2: ON/OFF input (5V/div).



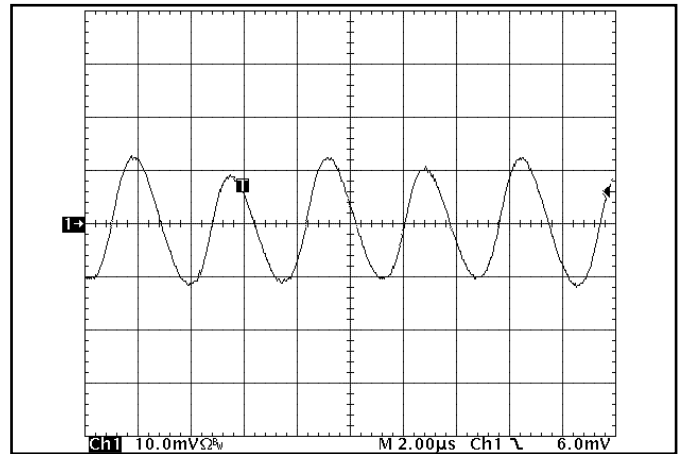
**Figure 11:** Output voltage response to step-change in load current (50%-75%-50% of  $I_{out(max)}$ ;  $dI/dt = 0.1A/\mu s$ ). Load cap: 10 $\mu F$ , 100 m $\Omega$  ESR tantalum cap and 1 $\mu F$  ceramic cap. Ch 1:  $V_{out}$  (50mV/div), Ch 2:  $I_{out}$  (20A/div).



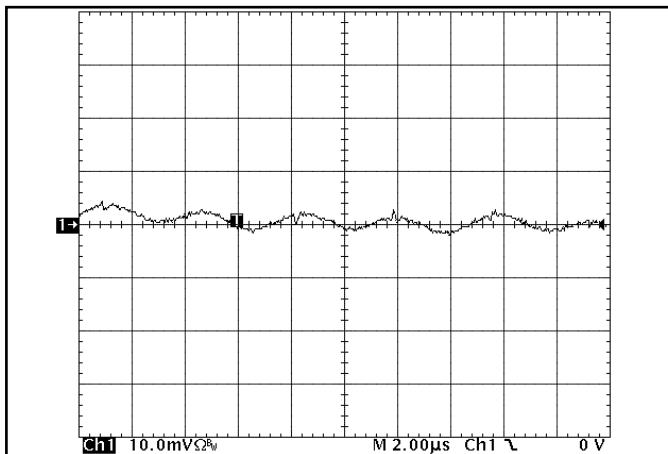
**Figure 12:** Output voltage response to step-change in load current (50%-75%-50% of  $I_{out(max)}$ ;  $dI/dt = 5A/\mu s$ ). Load cap: 470 $\mu F$ , 30 m $\Omega$  ESR tantalum cap and 1 $\mu F$  ceramic cap. Ch 1:  $V_{out}$  (100mV/div), Ch 2:  $I_{out}$  (20A/div).



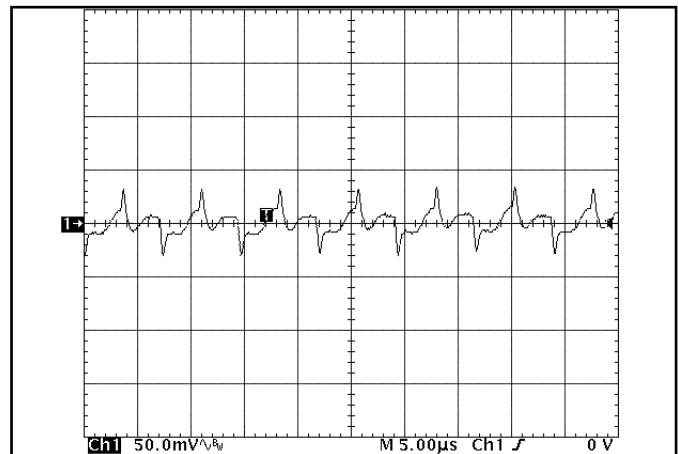
**Figure 13:** Test set-up diagram showing measurement points for Input Terminal Ripple Current (Figure 14), Input Reflected Ripple Current (Figure 15) and Output Voltage Ripple (Figure 16).



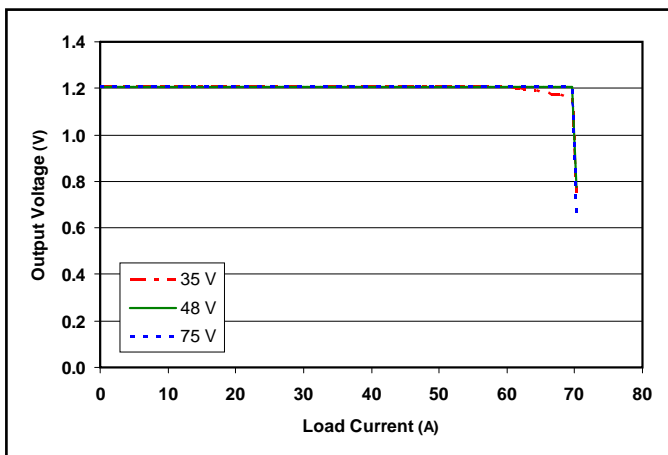
**Figure 14:** Input Terminal Ripple Current,  $i_c$ , at full rated output current and nominal input voltage with  $10\ \mu\text{H}$  source impedance and  $47\ \mu\text{F}$  electrolytic capacitor ( $100\ \text{mA}/\text{div}$ ). See Figure 13.



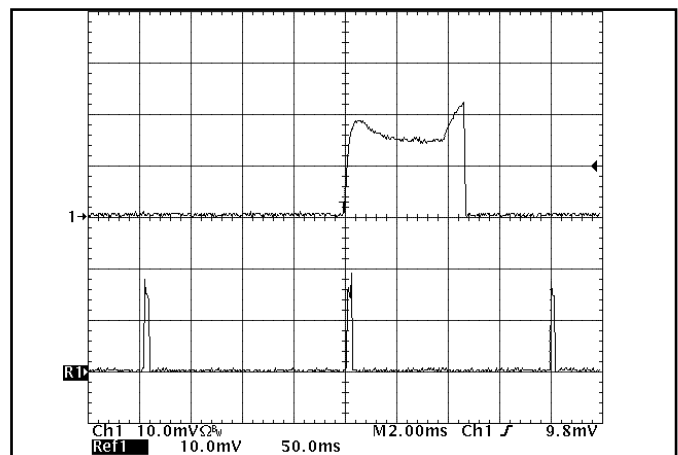
**Figure 15:** Input reflected ripple current,  $i_s$ , through a  $10\ \mu\text{H}$  source inductor at nominal input voltage and rated load current ( $10\ \text{mA}/\text{div}$ ). See Figure 13.



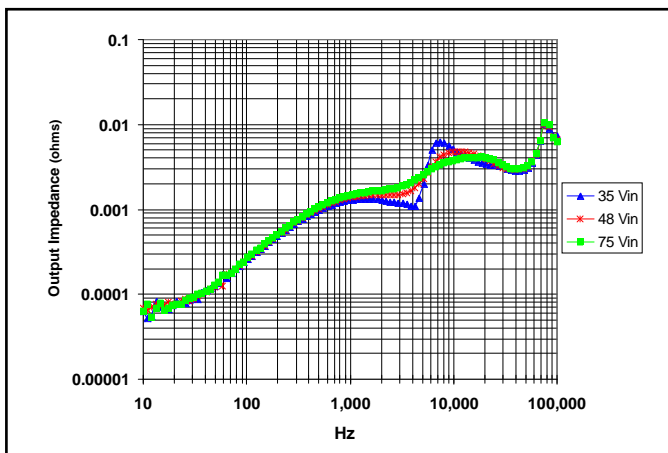
**Figure 16:** Output voltage ripple at nominal input voltage and rated load current ( $50\ \text{mV}/\text{div}$ ). Load capacitance:  $1\ \mu\text{F}$  ceramic capacitor and  $15\ \mu\text{F}$  tantalum capacitor. Bandwidth:  $500\ \text{MHz}$ . See Figure 13.



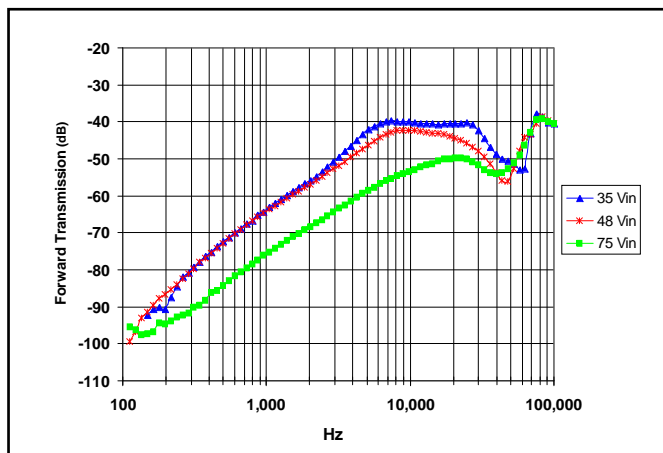
**Figure 17:** Output voltage vs. load current showing typical current limit curves and converter shutdown points.



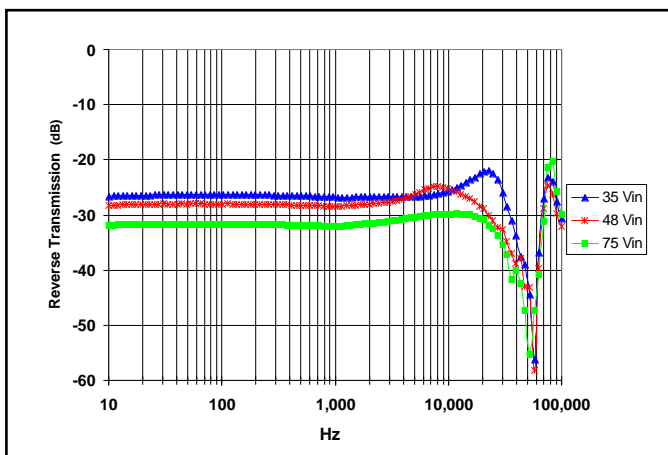
**Figure 18:** Load current ( $50\text{A}/\text{div}$ ) as a function of time when the converter attempts to turn on into a  $1\ \text{m}\Omega$  short circuit. Top trace ( $2\text{ms}/\text{div}$ ) is an expansion of the on-time portion of the bottom trace.



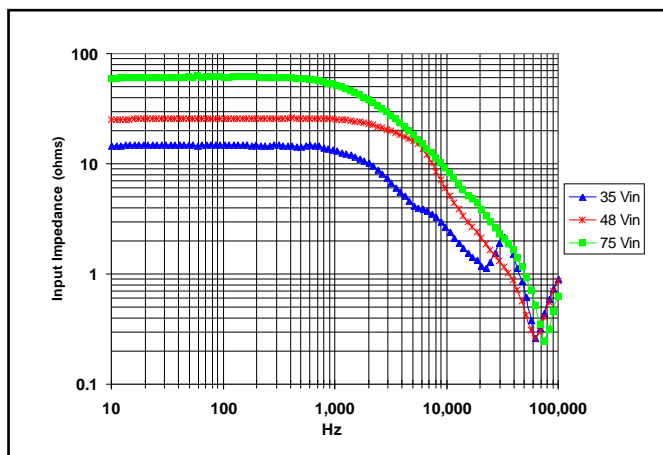
**Figure 19:** Magnitude of incremental output impedance ( $Z_{out} = v_{out}/i_{out}$ ) for minimum, nominal, and maximum input voltage at full rated power.



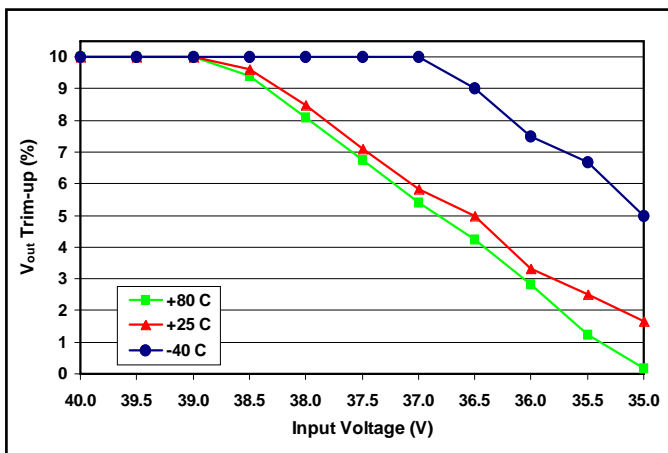
**Figure 20:** Magnitude of incremental forward transmission ( $FT = v_{out}/v_{in}$ ) for minimum, nominal, and maximum input voltage at full rated power.



**Figure 21:** Magnitude of incremental reverse transmission ( $RT = i_{in}/i_{out}$ ) for minimum, nominal, and maximum input voltage at full rated power.



**Figure 22:** Magnitude of incremental input impedance ( $Z_{in} = v_{in}/i_{in}$ ) for minimum, nominal, and maximum input voltage at full rated power.



**Figure 23:** Achievable trim-up percentage vs. input voltage. at +85°C, +25°C and -40°C (all at full load). Full trim-up percentage is achieved at loads of 50% or less across full temperature range.



### BASIC OPERATION AND FEATURES

The PowerQor series converter uses a two-stage power conversion topology. The first stage is a buck-converter that keeps the output voltage constant over variations in line, load, and temperature. The second stage uses a transformer to provide the functions of input/output isolation and voltage step-down to achieve the low output voltage required.

Both the first stage and the second stage switch at a fixed frequency for predictable EMI performance. Rectification of the transformer's output is accomplished with synchronous rectifiers. These devices, which are MOSFETs with a very low on-state resistance, dissipate far less energy than Schottky diodes. This is the primary reason that the PowerQor converter has such high efficiency, even at very low output voltages and very high output currents.

Dissipation throughout the converter is so low that it does not require a heatsink for operation. Since a heatsink is not required, the PowerQor converter does not need a metal baseplate or potting material to help conduct the dissipated energy to the heatsink. The PowerQor converter can thus be built more simply and reliably using high yield surface mount techniques on a PCB substrate.

The PowerQor series of half-brick, quarter-brick and eighth-brick converters uses the industry standard footprint and pin-out configuration.

### CONTROL FEATURES

**REMOTE ON/OFF (Pin 2):** The ON/OFF input, Pin 2, permits the user to control when the converter is on or off. This input is referenced to the return terminal of the input bus, Vin(-). There are two versions of the converter that differ by the sense of the logic used for the ON/OFF input.

In the positive logic version, the ON/OFF input is active high (meaning that a high turns the converter on). In the negative logic version, the ON/OFF signal is active low (meaning that a low turns the converter on). Figure A details five possible circuits for driving the ON/OFF pin. Figure B is a detailed look of the internal ON/OFF circuitry.

**REMOTE SENSE(+)** (Pins 8 and 6): The SENSE(±) inputs correct for voltage drops along the conductors that connect the converter's output pins to the load.

Pin 8 should be connected to Vout(+) and Pin 6 should be connected to Vout(-) at the point on the board where regulation is desired. A remote connection at the load can adjust for a voltage drop only as large as that specified in this datasheet, that is

$$[V_{out(+)} - V_{out(-)}] - [V_{sense(+)} - V_{sense(-)}] \leq \text{Sense Range \%} \times V_{out}$$

Pins 8 and 6 must be connected for proper regulation of the output voltage. If these connections are not made, the converter will deliver an output voltage that is slightly lower than its specified value.

Note: the output over-voltage protection circuit senses the voltage

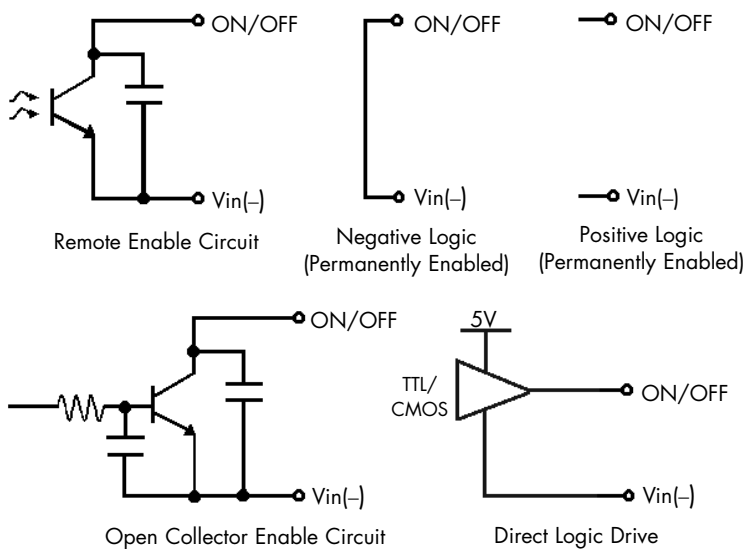


Figure A: Various circuits for driving the ON/OFF pin.

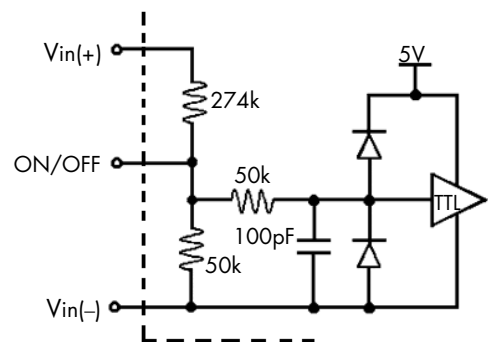


Figure B: Internal ON/OFF pin circuitry

across the output (pins 9 and 5) to determine when it should trigger, not the voltage across the converter's sense leads (pins 8 and 6). Therefore, the resistive drop on the board should be small enough so that output OVP does not trigger, even during load transients.

**OUTPUT VOLTAGE TRIM (Pin 7):** The TRIM input permits the user to adjust the output voltage across the sense leads up or down according to the trim range specifications.

To decrease the output voltage, the user should connect a resistor between Pin 7 and Pin 6 (SENSE(-) input). To increase the output voltage, the user should connect a resistor between Pin 7 and Pin 8 (SENSE(+) input). For a desired change of the nominal output voltage, the value of the resistor should be

$$R_{\text{trim}} = \left| \frac{V_{\text{DES}}}{V_{\text{NOM}} - V_{\text{DES}}} \right| - 1 \quad (\text{k}\Omega)$$

where

$V_{\text{NOM}}$  = Nominal Voltage

$V_{\text{DES}}$  = Desired Voltage

(Vertical bars indicate absolute value)

Figure C graphs the relationship between the trim resistor value and  $R_{\text{trim-up}}$  and  $R_{\text{trim-down}}$ , showing the total range the output voltage can be trimmed up or down.

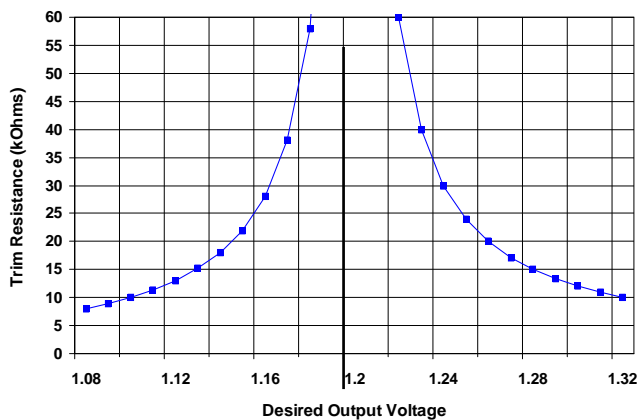


Figure C: Trim Graph for 1.2Vout module

**Note:** the TRIM feature does not affect the voltage at which the output over-voltage protection circuit is triggered. Trimming the output voltage too high may cause the over-voltage protection circuit to engage, particularly during transients.

It is not necessary for the user to add capacitance at the Trim pin. The node is internally bypassed to eliminate noise.

**Alternative Trim Equation:** When ordering this option (see ordering information page) Tyco compatible trim equations are applicable (range from -10% to +10%, see Figure D)

TRIM UP:

$$R_{\text{trim-up}} = \left[ \frac{9.769 \times V_{\text{NOM}} \times (100 + \Delta\%)}{0.6 \times \Delta\%} - \frac{1299.1}{\Delta\%} - 33.49 \right] (\text{k}\Omega)$$

TRIM DOWN:

$$R_{\text{trim-down}} = \left[ \frac{1299.1}{\Delta\%} - 33.49 \right] (\text{k}\Omega)$$

Where

$$\Delta\% = \left| \frac{V_{\text{DES}} - V_{\text{NOM}}}{V_{\text{NOM}}} \right| \times 100$$

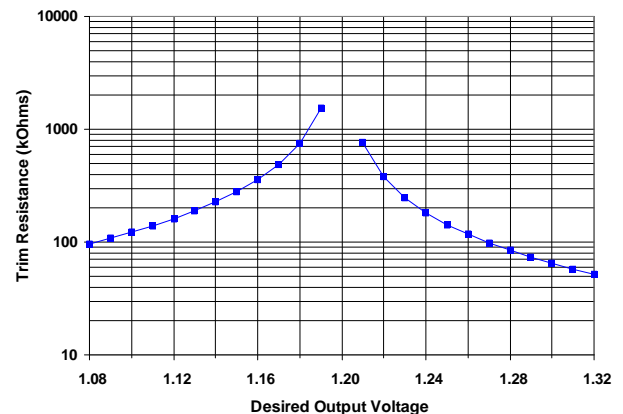


Figure D: Alternative Trim Graph for 1.2Vout module

**Total DC Variation of Vout:** For the converter to meet its full specifications, the maximum variation of the DC value of Vout, due to both trimming and remote load voltage drops, should not be greater than that specified for the output voltage trim range.

## PROTECTION FEATURES

**Input Under-Voltage Lockout:** The converter is designed to turn off when the input voltage is too low, helping avoid an input system instability problem, described in more detail in the application note titled "Input System Instability". The lockout circuitry is a comparator with DC hysteresis. When the input voltage is rising, it must exceed the typical Turn-On Voltage Threshold value (listed on the specification page) before the converter will turn

on. Once the converter is on, the input voltage must fall below the typical Turn-Off Voltage Threshold value before the converter will turn off.

**Output Current Limit:** The maximum current limit remains constant as the output voltage drops. However, once the impedance of the short across the output is small enough to make the output voltage drop below the specified Output DC Current-Limit Shutdown Voltage, the converter turns off.

The converter then enters a “hiccup mode” where it repeatedly turns on and off at a 5 Hz (nominal) frequency with a 5% duty cycle until the short circuit condition is removed. This prevents excessive heating of the converter or the load board.

**Output Over-Voltage Limit:** If the voltage across the output pins exceeds the Output Over-Voltage Protection threshold, the converter will immediately stop switching. This prevents damage to the load circuit due to 1) excessive series resistance in output current path from converter output pins to sense point, 2) a release of a short-circuit condition, or 3) a release of a current limit condition. Load capacitance determines exactly how high the output voltage will rise in response to these conditions. After 200 ms the converter will automatically restart.

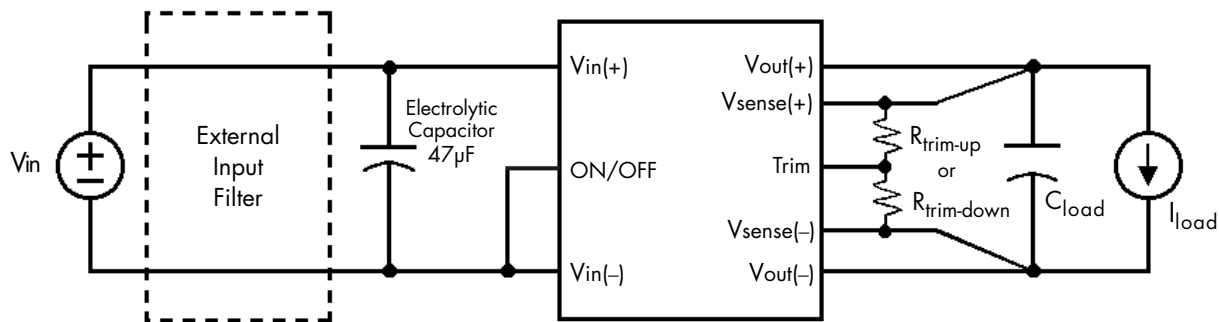
**Over-Temperature Shutdown:** A temperature sensor on the converter senses the average temperature of the module. The thermal shutdown circuit is designed to turn the converter off when the temperature at the sensed location reaches the Over-Temperature Shutdown value. It will allow the converter to turn on again when the temperature of the sensed location falls by the amount of the Over-Temperature Shutdown Restart Hysteresis value.

## APPLICATION CONSIDERATIONS

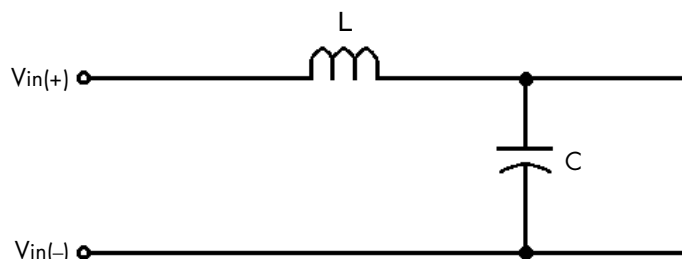
**Input System Instability:** This condition can occur because any DC/DC converter appears incrementally as a negative resistance load. A detailed application note titled “Input System Instability” is available on the SynQor web site ([www.synqor.com](http://www.synqor.com)) which provides an understanding of why this instability arises, and shows the preferred solution for correcting it.

**Application Circuits:** Figure E provides a typical circuit diagram which details the input filtering and voltage trimming.

**Input Filtering and External Capacitance:** Figure F provides a diagram showing the internal input filter components. This filter dramatically reduces input terminal ripple current,



**Figure E:** Typical application circuit (negative logic unit, permanently enabled).



**Figure F:** Internal Input Filter Diagram (component values listed on [specifications page](#)).

which otherwise could exceed the rating of an external electrolytic input capacitor. The recommended external input capacitance is specified in the "Input Characteristics" section. More detailed information is available in the application note titled "EMI Characteristics" on the SynQor website.

**Startup Inhibit Period:** The Startup Inhibit Period ensures that the converter will remain off for approximately 200ms when it is shut down for any reason. When an output short is present, this generates a 5Hz "hiccup mode," which prevents the converter from overheating. In all, there are seven ways that the converter can be shut down, initiating a Startup Inhibit Period:

- Input Under-Voltage Lockout
- Input Over-Voltage Shutdown (not present in Quarter-brick)
- Output Over-Voltage Protection
- Over Temperature Shutdown
- Current Limit
- Short Circuit Protection
- Turned off by the ON/OFF input

Figure G shows three turn-on scenarios, where a Startup Inhibit Period is initiated at  $t_0$ ,  $t_1$ , and  $t_2$ :

Before time  $t_0$ , when the input voltage is below the UVL threshold, the unit is disabled by the Input Under-Voltage Lockout feature. When the input voltage rises above the UVL threshold, the Input Under-Voltage Lockout is released, and a Startup Inhibit Period is initiated. At the end of this delay, the ON/OFF pin is evaluated, and since it is active, the unit turns on.

At time  $t_1$ , the unit is disabled by the ON/OFF pin, and it cannot be enabled again until the Startup Inhibit Period has elapsed.

When the ON/OFF pin goes high after  $t_2$ , the Startup Inhibit Period has elapsed, and the output turns on within the typical Turn-On Time.

**RoHS Compliance:** The EU led RoHS (Restriction of Hazardous Substances) Directive bans the use of Lead, Cadmium, Hexavalent Chromium, Mercury, Polybrominated Biphenyls (PBB), and Polybrominated Diphenyl Ether (PBDE) in Electrical and Electronic Equipment. This SynQor product is available as 5/6 RoHS compliant (product with lead) or 6/6 RoHS compliant. For more information please refer to SynQor's RoHS addendum available at our [RoHS Compliance / Lead Free Initiative](#) web page or e-mail us at [rohs@synqor.com](mailto:rohs@synqor.com).

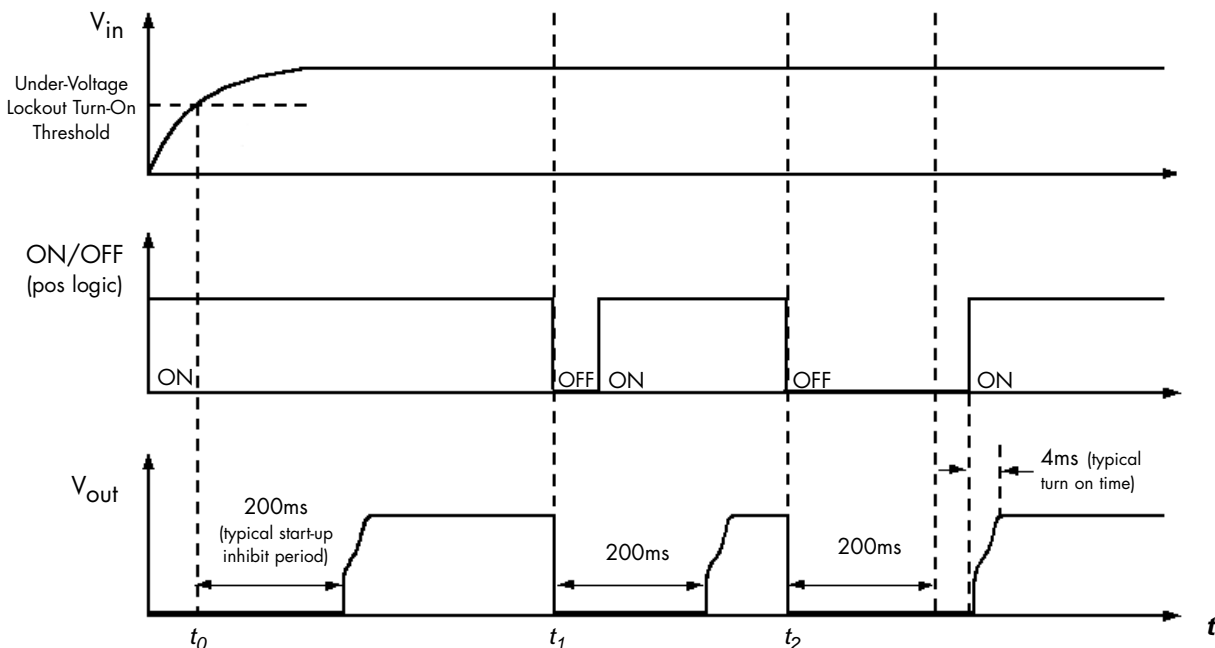


Figure G: Startup Inhibit Period (turn-on time not to scale)

